

# On Channel Estimators for Iterative CDMA Multiuser Receiver in Flat Rayleigh Fading

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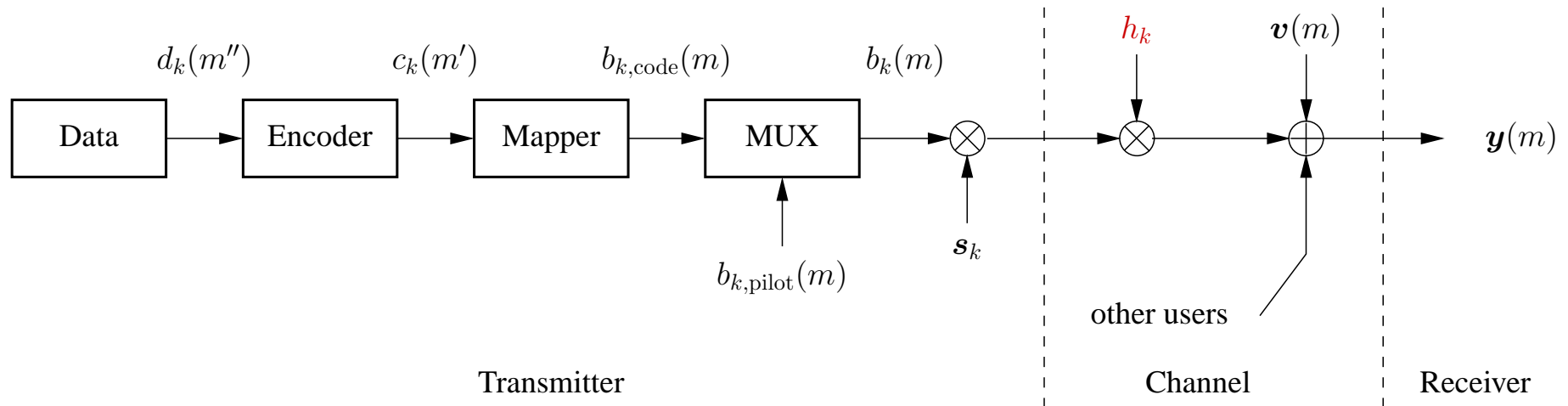
ICC 2004, Paris

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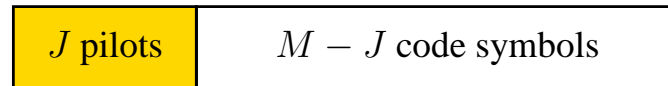
# Overview

1. System Setup
2. Detection & Decoding
3. Channel Estimators
4. Numerical Results
5. Conclusions

## Multuser Uplink



- Synch. CDMA system with **short** random spreading
- Convolutional encoder  $\mathcal{C}(5, 7)_8$
- QPSK Gray mapping
- $J$  pilots are placed in a preamble
- Single Rayleigh fading tap - **constant over  $M$  symbols**





## Signal Model for a Single Block

$$\begin{array}{c}
 \left[ \begin{array}{c} \text{---} \\ \text{---} \\ \vdots \\ \text{---} \end{array} \right] \\
 \mathbf{y} \\
 [MN \times 1]
 \end{array}
 =
 \begin{array}{c}
 \left[ \begin{array}{c} \left[ \begin{array}{c} | \\ | \\ | \\ | \end{array} \right] \\ \vdots \\ \left[ \begin{array}{c} | \\ | \\ | \\ | \end{array} \right] \end{array} \right] \\
 \mathbf{S} \\
 [MN \times MK]
 \end{array}
 \begin{array}{c}
 \left[ \begin{array}{c} \left[ \begin{array}{c} \diagdown \\ \diagup \end{array} \right] \\ \left[ \begin{array}{c} \diagdown \\ \diagup \end{array} \right] \\ \vdots \\ \left[ \begin{array}{c} \diagdown \\ \diagup \end{array} \right] \end{array} \right] \\
 \mathbf{B} \\
 [MK \times K]
 \end{array}
 \begin{array}{c}
 \left[ \begin{array}{c} \text{---} \\ \vdots \\ \text{---} \end{array} \right] \\
 \mathbf{h} \\
 [K \times 1]
 \end{array}
 +
 \begin{array}{c}
 \left[ \begin{array}{c} \text{---} \\ \vdots \\ \text{---} \end{array} \right] \\
 \mathbf{v} \\
 [MN \times 1]
 \end{array}
 \end{array}$$

$M$  block length,  $N$  spreading seq. length,  $K$  # users

## Multiuser Detector

### 1 - Parallel Interference Cancellation

$$\tilde{\mathbf{y}}_k^{(i)}(m) = \mathbf{y}(m) - \tilde{\mathbf{S}}^{(i)} \tilde{\mathbf{b}}^{(i)}(m) + \tilde{\mathbf{s}}_k^{(i)} \tilde{b}_k^{(i)}(m)$$

with  $\tilde{\mathbf{S}}^{(i)} = \mathbf{S} \text{diag}(\hat{\mathbf{h}}^{(i)})$ .

### 2 - LMMSE Filtering

$$\hat{b}_k^{(i+1)}(m) = \left( \mathbf{w}_k^{(i)} \right)^H \tilde{\mathbf{y}}_k^{(i)}(m)$$

The filter is the solution to

$$\mathbf{w}_k^{(i)}(m) = \underset{\mathbf{w}}{\text{argmin}} \underset{\mathbf{v}}{\mathbb{E}} \left\{ \left| b_k(m) - \mathbf{w}^H \tilde{\mathbf{y}}_k^{(i)}(m) \right|^2 \right\}.$$

## Decoder

The code symbol **a-posteriori probability** is defined as

$$\text{APP}_k(m') \triangleq \Pr \{c_k(m') = +1 | \mathbf{z}_k\}$$

and can be computed via the BCJR algorithm.

For a **Gaussian channel** the extrinsic probability and the a-posteriori probability satisfy the following relation

$$\text{APP}_k(m') \sim \text{EXT}_k(m') \frac{1}{\sqrt{2\pi\nu_k^2}} \exp\left(-\frac{|z_k(m') - \mu_k|^2}{2\nu_k^2}\right).$$

### Soft Symbol QPSK Mapping

$$\tilde{b}_{k,\text{code}}(m) = \frac{1}{\sqrt{2}} \{2\text{APP}_k(2m') - 1 + j[2\text{APP}_k(2m' + 1) - 1]\}$$

and similarly for extrinsic information.

## Approximated Concepts

### Linear signal model

$$\mathbf{y} = \mathbf{S}\mathbf{B}\mathbf{h} + \mathbf{v}$$

with  $\mathbf{h} \sim \mathcal{CN}(\mathbf{0}, \mathbf{I}_K)$  and  $\mathbf{v} \sim \mathcal{CN}(\mathbf{0}, \sigma_v^2 \mathbf{I}_N)$ .

**Approximation** Soft feedback data is treated as deterministic

$$\tilde{\mathbf{B}} = \begin{bmatrix} \mathbf{B}_{\text{Pilots}}; \mathbf{B}_{\text{Feedback}} \end{bmatrix}$$

### 1 - Approx. Least-Squares Estimator (ALS)

$$\hat{\mathbf{h}}_{\text{ALS}} = \left( \tilde{\mathbf{B}}^H \mathbf{S}^H \mathbf{S} \tilde{\mathbf{B}} \right)^{-1} \tilde{\mathbf{B}}^H \mathbf{S}^H \mathbf{y}$$

### 2 - Approx. linear MMSE Estimator (ALMMSE)

$$\hat{\mathbf{h}}_{\text{ALMMSE}} = \left( \tilde{\mathbf{B}}^H \mathbf{S}^H \mathbf{S} \tilde{\mathbf{B}} + \sigma_v^2 \mathbf{I}_K \right)^{-1} \tilde{\mathbf{B}}^H \mathbf{S}^H \mathbf{y}$$

## Linear MMSE (LMMSE) Concept

### 3 - Improved Estimator Considering Symbol Reliability

$$\hat{\mathbf{h}}_{\text{LMMSE}} = \mathbf{C}_{hy} \mathbf{C}_{yy}^{-1} \mathbf{y}$$

$$\mathbf{C}_{yh} = \mathbb{E}_{\mathcal{B}, \mathbf{h}, \mathbf{v}} \{ \mathbf{y} \mathbf{h}^H \} = \mathbf{S} \tilde{\mathbf{B}} \mathbf{T}$$

$$\mathbf{C}_{yy} = \mathbb{E}_{\mathcal{B}, \mathbf{h}, \mathbf{v}} \{ \mathbf{y} \mathbf{y}^H \} = \mathbf{S} \mathbb{E}_{\mathcal{B}} \{ \mathbf{B} \mathbf{T} \mathbf{B}^H \} \mathbf{S}^H + \sigma_v^2 \mathbf{I}_{KM}.$$

Statistical independence among code symbols leads to

$$\mathbb{E} \{ b_p(m) b_q^*(n) \} = \begin{cases} \tilde{b}_p(m) \tilde{b}_q^*(n) & \text{for } p \neq q, m \neq n \\ 1 & \text{for } p = q, m = n. \end{cases}$$

$\mathbf{T}$  is the correlation matrix of the channel taps

$$\mathbf{T} = \text{diag} (\sigma_{h,1}^2, \sigma_{h,2}^2, \dots, \sigma_{h,K}^2).$$

## Simulation Parameters

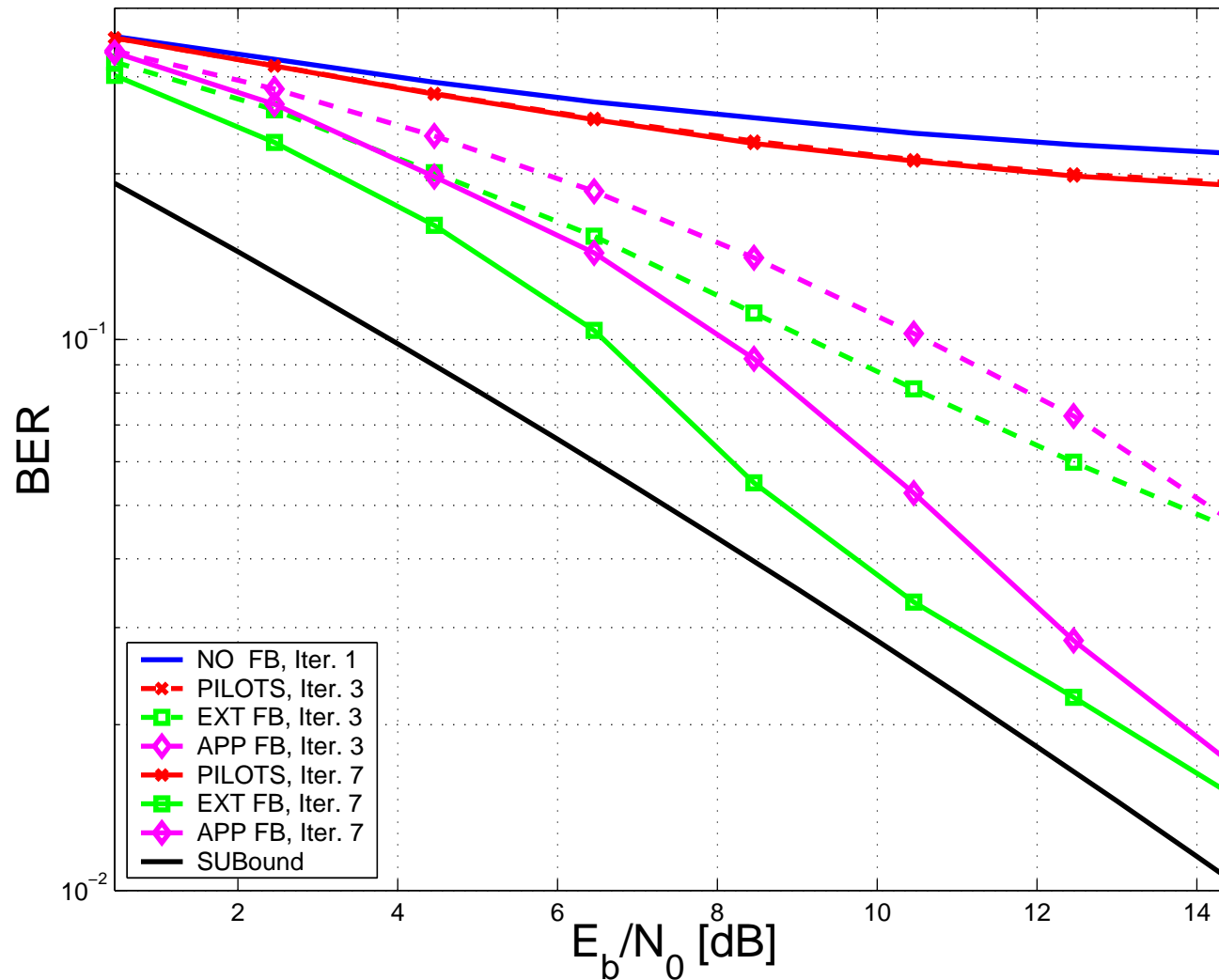
- Spreading sequence length  $N = 8$
- Number of users  $K = 12$ , **overloaded system**  $\alpha = 1.5$
- Block length  $M = 100$
- Number of pilots  $J = 10$
  
- Energy per information bit to noise level ratio

$$E_b/N_0 \triangleq \frac{1}{\sigma_v^2} \frac{M}{M-J}$$

- Single-User Bound (SUBound)

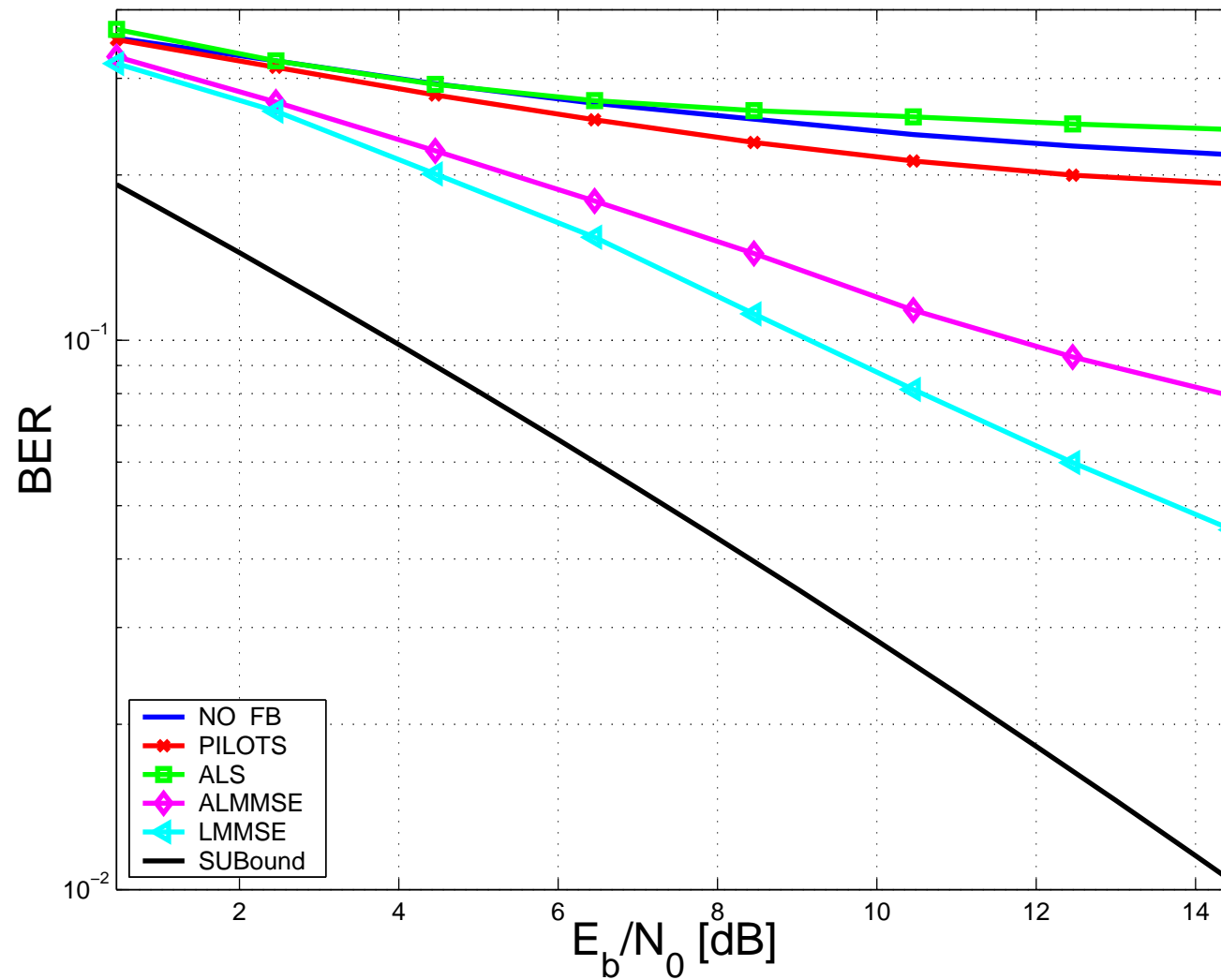
Performance of one user with perfect channel knowledge.

## APP vs. EXT in the Feedback

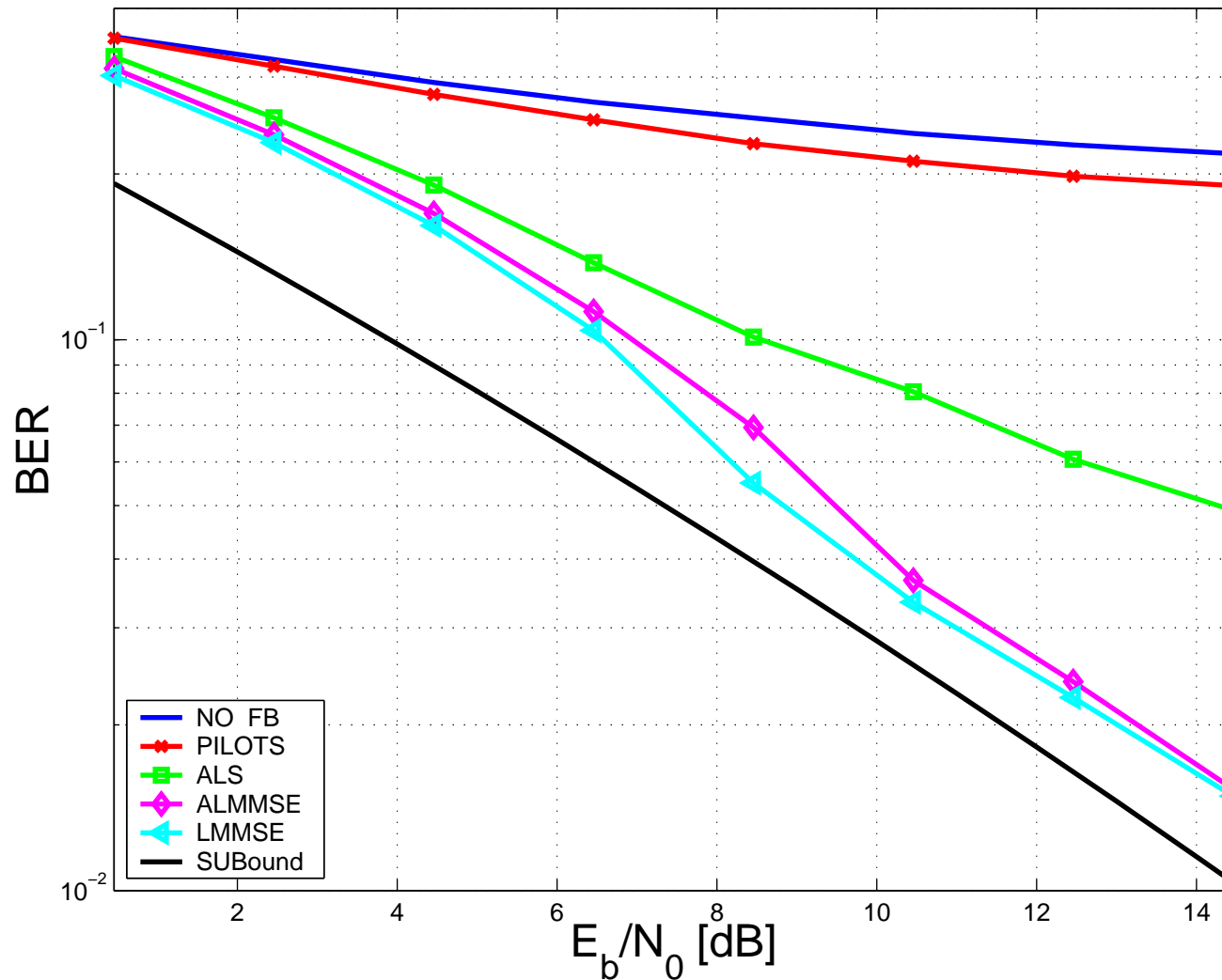


Extrinsic FB  
better than  
APP FB for  
channel  
estimation.

## BER after 3rd Iteration



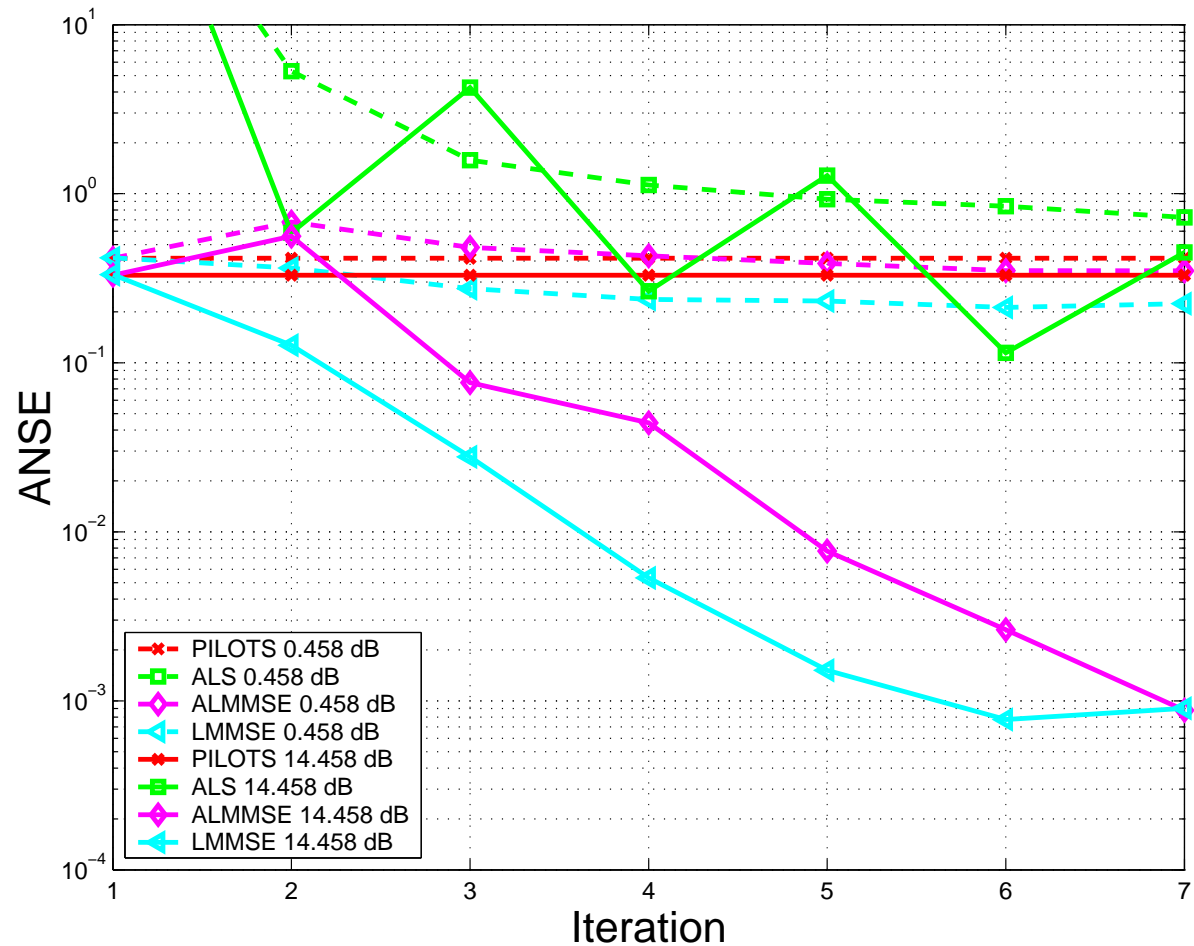
## BER after 7th Iteration



## Averaged Normalized Squared Error

Realizations  $R = 100$

$$\text{ANSE} \triangleq \frac{1}{R} \sum_{i=1}^R \frac{\|\mathbf{h} - \hat{\mathbf{h}}\|^2}{\|\mathbf{h}\|^2}$$



## Conclusions

- Soft symbols in feedback substantially improve channel estimation and overall receiver performance.
- Extrinsic info shows better performance than APP info in feedback.
- Improved LMMSE estimator “weights” reliability best.
- ANSE decreases monotonically.